Heating Effects in Nanoscale Devices

Dragica Vasileska¹, Katerina Raleva² and Stephen M. Goodnick¹

¹*Arizona State University, Tempe, AZ, 85287-5706, USA*

²*University Sts Cyril and Methodius, Skopje, Macedonia*

1. Introduction

The ever increasing demand for faster microprocessors and the continuous trend to pack more transistors on a single chip has resulted in an unprecedented level of power dissipation, and therefore higher temperatures at the chip level. Thermal phenomena are not directly responsible for the electrical functionality and performance of semiconductor devices, but adversely affect their reliability. Four major thermally-induced reliability concerns for transistors are: (1) degradation of device thermal characteristics due to heating effects, (2) failure due to the electrostatic discharge phenomenon, (3) stresses due to different rates of thermal expansion of transistor constituents, and (4) failure of metallic interconnects due to diffusion or flow of atoms along a metal interconnect in the presence of a bias current, known as the electromigration phenomenon. Self-heating of the device and interconnects reduces electron mobility and results in a poor or, at best, non-optimal, performance of these devices and structures. Fig. 1 shows the trend of the average power density for high-performance microprocessors according to the ITRS. No flattening or slower decelerated increase will occur after the introduction of the Silicon on Insulator (SOI) technology. It should be noted that the power density shown in Fig. 1 is the average power density, i.e. the total chip power divided by the chip area. In logic circuits, such as microprocessors, the power is non-uniformly distributed. There are portions of the chip of quite low power dissipation (memory blocks) and, on the other hand, portions running at full speed with high activity factors where the power density can easily be more than a magnitude higher than the average chip power density from Fig. 1. The latter portions will create hot spots with quite high local temperature. The power density in the active transistor region (essentially the channel region underneath the gate) is again much higher than the average power density in a hot spot when the transistor is in the on-state. Thus, the treatment of self-heating and the realistic estimation of the power density is quite a complex problem.

Sometime within the next five years, traditional CMOS technology is expected to reach limits of scaling. As channel lengths shrink below 22 nm, complex channel profiles are required to achieve desired threshold voltages and to alleviate the short-channel effects. To fabricate devices beyond current scaling limits, Integrated Circuits (IC) companies are simultaneously pushing planar, bulk silicon CMOS design while exploring alternative gate stack materials (high-k dielectrics and metal gates), band engineering methods (using

strained Si or SiGe), and alternative transistor structures, such as fully-depleted (FD) SOI devices, dual gate (DG) structures, and FinFETs.

The problem with the SOI devices is that they exhibit self-heating effects (Majumdar, 1993; Pop, 2006). The buried oxide layer underneath the thin active silicon layer, having a thermal conductivity 100 times smaller than the value of bulk Si thermal conductivity, constitutes a large barrier for the heat removal from the active devices to the heat sink. Thus, it will become much more difficult to remove the heat generated in the active devices and considerable self-heating is expected to occur in the SOI transistors. This, in turn, causes a decrease of the on-currents and an increase of the off-current leading to a dramatic deterioration of transistor switching performance with consequences on the overall circuit performance. Also, the thermal conductivity of the thin silicon film decreases due to phonon boundary scattering.

Note that the removal of heat from deep inside the structure becomes very important issue in Today's nano-electronics Industry. Various novel semiconductor thermoelectric coolers and structures, such as thermionic (Shakouri *et al.*, 1998) and nanowire (Chen & Shakouri, 2002) coolers have been proposed, developed, and investigated due to the advance of nanotechnology. For example, thermionic emission current in heterostructures can be used to achieve evaporative cooling by selective emission of hot electrons over a barrier layer from cathode to anode. Such structure can effectively build a temperature gradient within the range of electron mean free path (a few hundred nanometers), which can be used to remove the heat from a CMOS hotspot.

Fig. 1. Evolution of the average power density in microprocessors in the past and expectations of the ITRS until 2018.

In summary, the self-heating effect is particularly important for transistors in SOI circuitry where the device is separated from the substrate by a low thermal conductivity buried silicon dioxide layer as well as copper interconnects that are surrounded by low thermal conductivity dielectric materials (Borkar, 1999). Accurate thermal modeling and design of

microelectronic devices and thin film structures at micro- and nanoscales poses a challenge to the thermal engineers who are less familiar with the basic concepts and ideas in sub-continuum heat transport.

2. Some General Aspects of Heat Conduction

The energy given up by the constituent particles such as atoms, molecules, or free electrons of the hotter regions of a body to those in cooler regions is called heat. Conduction is the mode of heat transfer in which energy exchange takes place in solids or in fluids in rest (i.e. no convection motion resulting from the displacement of the macroscopic portion of the medium) from the region of high temperature to the region of low temperature due to the presence of temperature gradient in the body. The heat flow can not be measured directly, but the concept has physical meaning because it is related to the measurable scalar quantity called temperature. Therefore, once the temperature distribution $T(r,t)$ within a body is determined as a function of position and time, then the heat flow in the body is readily computed from the laws relating heat flow to the temperature gradient. The science of heat conduction is principally concerned with the determination of temperature distribution within solids.

The basic law that gives the relationship between the heat flow and the temperature gradient, based on experimental observations, is generally named after the French mathematical physicist Joseph Fourier, who used it in his analytic theory of heat. For a homogeneous, isotropic solid (i.e. material in which thermal conductivity is independent of direction) the Fourier law is given in the form

$$q(r,t) = -\kappa \nabla T(r,t) \quad W/m^2 \quad (1)$$

where the temperature gradient is a vector normal to the isothermal surface, the heat flux vector $q(r,t)$ represents heat flow per unit time, per unit area of the isothermal surface in the direction of the decreasing temperature, and κ is called the thermal conductivity of the material which is a positive, scalar quantity. Since the heat flux vector $q(r,t)$ points in the direction of decreasing temperature, a minus sign is included in Eq. (1) to make the heat flow a positive quantity. When the heat flux is in W/m^2 and the temperature gradient is in $^{\circ}C/m$, the thermal conductivity κ has the units $W/(m^{\circ}C)$.

Clearly, the heat flow rate for a given temperature gradient is directly proportional to the thermal conductivity κ of the material. Therefore, in the analysis of heat conduction, the thermal conductivity of the material is an important property, which controls the rate of heat flow in the medium. There is a wide difference in the thermal conductivities of various engineering materials. The highest value is given by pure metals and the lowest value by gases and vapors; the amorphous insulating materials and inorganic liquids have thermal conductivities that lie in between. Thermal conductivity also varies with temperature. For most pure metals it decreases with temperature, whereas for gases it increases with increasing temperature.

At nanometer length scales, the familiar continuum Fourier law for heat conduction is expected to fail due to both classical and quantum size effects (Geppert, 1999; Zeng *et al.*, 2003; Majumdar, 1993). The past two decades have seen increasing attention to thermal conductivity and heat conduction in nanostructures. Experimental methods for characterizing the thermal conductivity of thin films and nanowires have been developed

and are still evolving. Experimental data have been reported on various nanostructures: thin films, superlattices, nanowires, and nanotubes. Along the way, models and simulations have been developed to explain the experimental data. This section summarizes some past work and the current understanding of heat conduction in nanostructures. We first give a brief overview on the fundamental physics that distinguishes phonon heat conduction in nanostructures from that in macrostructures. Then we discuss a few size effects in nanostructures that impact their thermal conductivity.

Heat conduction in dielectric materials and most semiconductors is dominated by lattice vibrational waves. The basic energy quantum of lattice vibration is called a phonon, analogous to a photon which is the basic energy quantum of an electromagnetic wave. Similar to photons, phonons can be treated as both waves and particles. Size effects appear if the structure characteristic length is comparable to or smaller than the phonon characteristic length. Two kinds of size effects can exist: the classical size effect, when phonons can be treated as particles, and the wave effect, when the wave phase information of phonons becomes important. Distinction between these two regimes depends on several characteristic lengths (Chen, 2001). The important characteristic lengths of phonon heat conduction are the mean free path, the wavelength, and the phase coherence length (Reif, 1985). The mean free path is the average distance that phonons travel between successive collisions. The mean free path Λ is often estimated from kinetic theory as

$$\kappa = \frac{1}{3} C_V v_p \Lambda \quad (2)$$

where C_V and v_p are the volumetric specific heat capacity of a phonon, and the phonon velocity inherent in a material. In silicon, for example, the phonon mean free path is on the order of ~ 300 nm (Ashegi *et al.*, 1998). The phase of a wave can be destroyed during collision, which is typically the case in inelastic scattering processes, such as the phonon-phonon collision.

Fig. 2. Regime map for phonon transport in ultra-thin silicon layers. Mean free path Λ is a distance that phonons travel on average before being scattered by other phonons. If the dimension of the silicon layer is smaller than Λ , the Boltzmann Transport Equation (BTE) should be used for heat transfer analysis of the thin film. The dominant phonon wavelength, λ , at room temperature, is on the order of 2-3 nm. Analogous phonon wave simulations should be performed for devices with thicknesses comparable to λ .

Fig. 3. Phonon-boundary scattering is responsible for a large reduction in the thermal conductivity of a thin silicon layer where the thickness of the film, d_s , is comparable to or smaller than the phonon mean free path, Λ .

Fig. 2 compares the dimensions of several nanostructures (e.g., an SOI device and a superlattice structure) with the dominant phonon mean free path (MFP) and wavelength at room temperature. This graph also provides a general guideline for the appropriate treatment of phonon transport in nanostructures. Phonon transport can be predicted using the Boltzmann particle transport equations (BTE), which is required only when the scattering rates of electrons or phonons vary significantly within a distance comparable to their respective mean free paths.

The BTE simply takes care of the bin counting of the energy carrier particles of a given velocity and momentum, scattering in and out of a control volume at a point space and time. Analysis of the heat transfer in microelectronic devices, interconnects and nanostructures using the BTE is very cumbersome and complicated, even for simple geometries, and has been the topic of research and development in the field of micro- and nanoscale heat transfer for the past two decades (Choi & Maruyama, 2003). Eq. (2) can only be considered as providing the qualitative behavior of a thermal conductivity from which the thermal conductivity is found to be proportional to the phonon mean free path (MFP). The phonon MFP is well known to become shorter as the system is hotter because the phonon population is increased, which causes the collision frequency among phonons to be high. Increased phonon collisions prevent the phonons with high energy in the hot region from moving to the cold region and vice versa. This means that the energy transport is low; consequently the thermal conductivity is low. Therefore, it can be inferred that phonon scattering governs the thermal conductivity.

Detailed descriptions and analyses of the ballistic heat transfer in a semiconductor/ metallic layer are beyond the scope of this book chapter. However, the most prominent manifestation of ballistic heat transport in thin films would occur in the form of large reductions in thermal conductivity compared to the bulk values. Ballistic phonon transport in silicon films, or phonon-boundary scattering (see Fig. 3), has been investigated through large measured reductions in the lateral thermal conductivity compared to the bulk value near room temperature (Ju & Goodson, 1995; Liu & Asheghi, 2004; ^aLiu & Asheghi, 2005).

The lateral thermal conductivity of the thin silicon layer decreases as the thickness of the film is reduced. Deviation of the thermal conductivity from the bulk value takes a sharp dive as the thickness of the film is reduced beyond 300 nm, which is the order of magnitude for the phonon mean free path in silicon at room temperature. For example, the thermal conductivity of the 20 nm thick silicon layer is nearly an order of magnitude smaller than the bulk value. The impact of phonon-boundary scattering on the thermal conductivity of a thin silicon layer can be predicted using the BTE solution and the theory described by Asheghi *et al.*, such that it agrees very well with the experimental data.

One way to estimate the impact of the micro/nanoscale effect is to use the modified thermal conductivity values for thin silicon and copper layers in conventional thermal simulation tools that use the continuum theory or diffusion equation. In order to perform more realistic estimates of the current degradation and the hot spot temperature we have followed the approach of Sondheimer (Sondheimer, 2001) that takes into account phonon boundary scattering by assuming it to be purely diffusive. Namely, the thermal conductivity of a semiconductor film of a thickness a , under the assumption that the z -axis is perpendicular to the plane of the film, the surfaces of the film being at $z=0$ and $z=a$, is given by:

$$\kappa(z) = \kappa_0(T) \int_0^{\pi/2} \sin^3 \theta \left[1 - \exp\left(-\frac{a}{2\lambda(T)\cos\theta}\right) \cosh\left(\frac{a-2z}{2\lambda(T)\cos\theta}\right) \right] d\theta \quad (3)$$

where $\lambda(T)$ is the mean free path expressed as $\lambda(T) = \lambda_0(300/T)$ nm where room temperature mean free path of bulk phonons is taken to be $\lambda_0 = 290$ nm. Selberherr (Palankovski & Selberherr, 2001; Sivaco Manuals) has parameterized the temperature dependence (Asheghi *et al.*, 1998) of the bulk thermal conductivity in the temperature range between 250K and 1000K. In our case we find that the appropriate expression is:

$$\kappa_0(T) = \frac{135}{a + bT + cT^2} \quad \text{W/m/K} \quad (4)$$

where $a=0.03$, $b=1.56 \times 10^{-3}$, and $c=1.65 \times 10^{-6}$. Eqs. (3) and (4) give excellent agreement with the experimental and the theoretical data reported in a later Asheghi paper (Asheghi *et al.*, 2005). Our group has utilized Eqs. (3) and (4) to obtain both the temperature and thickness dependence of the thermal conductivity for different temperatures, as shown in Fig. 4, compared to experimental data of Asheghi and co-workers (Asheghi *et al.*, 2005) at 300K. This model for the thermal conductivity is then implemented into the energy balance equation for acoustic phonons in our electro-thermal device simulator, and different device technology nodes have been examined.

Fig. 4. Silicon film thickness dependence of the average thermal conductivity at $T=300$ K vs. active silicon layer thickness. Experimental data are taken from the work of Asheghi and co-workers (Asheghi *et al.*, 2005).

In the ultra-fast laser heating processes at time scales of 10^{-15} to 10^{-12} seconds, as well as high speed transistors switching at timescales in the order of 10^{-11} seconds, the temperatures of the electron and phonon systems are not in equilibrium and may differ by orders of magnitude. Even after the phonon and electron reach equilibrium, the energy carried away by phonons can travel only to 10-100 nanometers; therefore, the temperature of the transistor can easily rise to several times its designed reliability limit. Under these circumstances, regardless of the cooling solution at the packaging level, a catastrophic failure at the device level can occur, because the impact of the rapid temperature rise is limited to the device and its vicinity. As a result, while the package level cooling solutions can reduce the quasi steady-state/average temperature across a microprocessor or at length scales in the order of one millimeter, it has very little impact at micro/nanoscales. Basically, there is no practical way to reduce the temperature at the device and interconnect level by means of a cooling device or solution; therefore, the options for thermal engineering of these devices are very limited. However, intelligent electro-thermal design along with careful floor planning at the device level can largely reduce the temperature rise within a device. This means that the role of the thermal engineer is to properly anticipate – perhaps in full collaboration with electrical engineers – and prevent the problem at the early stages and at the device level, rather than to pass the problem to the package-level thermal engineers.

3. Early attempts to Modeling Heating Effects in State of the Art Devices

In order to understand the modeling of heating effects in commercial device simulators, we first derive the differential equation of heat conduction for a stationary, homogeneous, isotropic solid with heat generation within the body. Heat generation in general may be due to nuclear, electrical, chemical, or other sources that may be a function of time and/or position. The heat generation rate in the medium, generally specified as heat generation per unit time, per unit volume, is denoted $H(r,t)$ and is given in W/m^3 .

We consider the energy balance equation for a small control volume stated as: rate of heat entering through the bounding surfaces of V + rate of energy generation in V = rate of storage of energy in V . In other words the rate of heat entering through the bounding surfaces of V (term 1) is:

$$Term1 = -\int_A \mathbf{q} \cdot \mathbf{n} dA = -\int_V \nabla \cdot \mathbf{q} dV \quad (5)$$

where A is the surface area of the volume element V , $\mathbf{n}$ is the outward-drawn normal unit vector to the surface element dA , $\mathbf{q}$ is the heat flux vector at dA ; here the minus sign is included to ensure that the heat flow is into the volume element V , and the divergence theorem is used to convert the surface integral to volume integral. The remaining two terms are evaluated as:

$$Term2 = \text{Rate of energy generation in } V = \int_V H(r,t) dV \quad (6)$$

$$Term3 = \text{rate of energy storage in } V = \int_V \rho C_v \frac{\partial T(r,t)}{\partial t} dV \quad (7)$$

Combining Eqs. (5) – (7) yields:

$$\int_V \left[-\nabla \cdot q(r,t) + H(r,t) - \rho C_V \frac{\partial T(r,t)}{\partial t} \right] dV = 0 \quad (8)$$

The last equation is derived for an arbitrary small volume element V within the solid, hence the volume V may be chosen so small as to remove the integral and one obtains:

$$-\nabla \cdot q(r,t) + H(r,t) = \rho C_V \frac{\partial T(r,t)}{\partial t} \quad (9)$$

Substituting $q(r,t)$ from Eq. (1) into Eq. (9) finally yields the differential equation of heat conduction for a stationary, homogeneous, isotropic solid with heat generation within the body as:

$$-\nabla \cdot [k \nabla T(r,t)] + H(r,t) = \rho C_V \frac{\partial T(r,t)}{\partial t} = c \frac{\partial T(r,t)}{\partial t} \quad (10)$$

The energy generation term in Eq. (10) is discussed in more details in Section 3.1 below.

3.1 Form of the Heat Source Term

Lai and Majumdar (Lai & Majumdar, 1996) developed a coupled electro-thermal model for studying thermal non-equilibrium in submicron silicon MOSFETs. Their results showed that the highest electron and lattice temperatures occur under the drain side of the gate electrode, which, also corresponds to the region where non-equilibrium effects such as impact ionization and velocity overshoot are maximum. Majumdar *et al.* (Majumdar *et al.*, 1995) have analyzed the variation of hot electrons and associated hot phonon effects in GaAs MESFETs. These hot carriers were observed to decrease the output drain current by as much as 15%. Thus, they concluded that both electron and lattice heating should be included in the electrical behavior of devices.

As it has been recognized that the simulation of devices operated under non-isothermal conditions was of growing importance, the heat flow equation given in Eq. (10) has been added to conventional drift-diffusion and/or hydrodynamic models to account for the mobility degradation due to lattice heating. There has been a discussion on the form of the heat generation term, details of which can be found in an excellent paper by Wachutka (Wachutka, 1990). Briefly, three different models are most commonly used and these include: (1) Joule Heating, (2) electron-lattice scattering and (3) the phonon model. Although these three models yield identical results in equilibrium, under non-equilibrium conditions the results of the three models can vary significantly.

Case 1: Within the Joule heating model, the thermal model consists of the heat diffusion equation using a Joule heating term as the source. The source term is computed from the electrical solution as the product of the local field and the current density (Gaur & Navon, 1976)

$$H = \mathbf{J} \cdot \mathbf{E} \quad (11)$$

This source term is similar to the one used by Leung and co-workers (Leung *et al.*, 1997) and assumes that recombination heating is negligible. In this case the “hot spot” will occur near the location where the dot product of the field and of the current density is the largest. Simulations that have used this expression as a heating term suggest that the bulk of the heating will occur directly under the gate region where most of the voltage drop occurs and where the current density is the largest because of the restricted electron flow path due to the depletion regions. A study by Raman and co-workers on lightly doped drain (LDD) devices suggested that the location of the hot-spot occurs at the drain side of the gate. The complete Leung and co-workers expression used for the source term is

$$H = \mathbf{J} \cdot \mathbf{E} + (R - G)(E_G + 3k_B T) \quad (12)$$

where the second term represents the heating rate due to non-radiative generation (G) and recombination (R) of electron-hole pairs. E_G is the semiconductor band-gap, k_B is the Boltzmann constant and T is the lattice temperature.

Case 2: Within the electron-lattice scattering model, the thermal system is represented as a single lattice temperature and is considered to be in thermal equilibrium. However, since the heat generation is due to non-equilibrium electron temperatures, the source term is then taken as a scattering term obtained from the relaxation time approximation and moments of the BTE. In essence, the transport is similar to case 1 in that the heat diffusion equation governs transport in the solid, except for the fact that the source term is now given as a moment of the relaxation time approximation, i.e.

$$H = \frac{3\rho k_B}{2} \left(\frac{T_e - T_L}{\tau_{e-L}} \right) \quad (13)$$

In the above expression T_e is the electron temperature, T_L is the lattice temperature and τ_{e-L} is the electron lattice scattering time constant.

Case 3: Phonon-model. Under thermal non-equilibrium conditions a system of two phonons is used as represented later in the text. In this case, the ‘lattice’ temperature is taken to be the acoustic phonon temperature T_A , because this is the mode responsible for diffusion. The energy balance equations for the acoustic and optical modes were for the first time derived by Majumdar and co-workers starting from the phonons Boltzmann transport equation. In all our investigations we have pursued this approach for the description of the phonon bath. A variant of this approach, that has been pursued by the Leeds group (Sadi *et al.*, 2007) and by Eric Pop and co-workers [Pop, 2006], counts the number of generated acoustic and optical phonons in a given branch and mode. Then, the total heat generation rate per unit volume is computed as

$$H = \frac{n}{N_{sim} \Delta t} \sum (\hbar \omega_{ems} - \hbar \omega_{abs}) \quad (14)$$

where n is the electron density, N_{sim} is the number of simulated particles and Δt is the simulation time.

In our opinion, the best solution to the submicron heat transport problem only has been provided by Narumanchi and co-workers (Naramunchi, 2004). In their recent study they propose a model based on the solution of the BTE, accounting for the transverse acoustic and longitudinal acoustic as well as optical phonons. Their model incorporates realistic phonon dispersion curves for silicon. The interactions among the different phonon branches and different phonon frequencies are considered and the proposed model satisfies energy conservation. Frequency-dependent relaxation times, obtained from perturbation theory, and accounting for phonon interaction rules, are used. In the calculation of the relaxation rates, they have included impurity scattering and the three-phonon interactions (the normal (N) process and the Umklapp (U) process). U processes pose direct thermal resistance while N processes influence the thermal resistance by altering the phonon distribution function. In this study, the BTE is numerically solved using a structured finite volume approach. Using this model, experimental in-plane thermal conductivity data for silicon thin films over a wide range of temperatures are matched satisfactorily.

4. The ASU Approach to Solving Lattice Heating Problem in Nanoscale Devices

To properly treat heating without any approximations made in the problem at hand, one in principle has to solve the coupled Boltzmann transport equations for the electron and phonon systems together. More precisely, *one has to solve the coupled electron – optical phonons – acoustic phonons – heat bath problem*, where each sub-process involves different time scales and has to be addressed in a somewhat individual manner and included in the global picture via a self-consistent loop. Let us consider the coupled system of semi-classical Boltzmann transport equations for the distribution functions of electrons $f(\mathbf{k}, \mathbf{r}, t)$ and phonons $g(\mathbf{k}, \mathbf{r}, t)$:

$$\left(\frac{\partial}{\partial t} + v_e(\mathbf{k}) \cdot \nabla_r + \frac{e}{\hbar} E(\mathbf{r}) \cdot \nabla_k \right) f = \sum_q \left\{ W_{e,q}^{k+q \rightarrow k} + W_{a,-q}^{k+q \rightarrow k} - W_{e,-q}^{k \rightarrow k+q} - W_{a,q}^{k \rightarrow k+q} \right\} \quad (16)$$

$$\left(\frac{\partial}{\partial t} + v_p(q) \cdot \nabla_r \right) g = \sum_k \left\{ W_{e,q}^{k+q \rightarrow k} - W_{a,q}^{k \rightarrow k+q} \right\} + \left(\frac{\partial g}{\partial t} \right)_{p-p}$$

Here $W_{e,q}^{k+q \rightarrow k}$ is the probability for electron transition from $k+q$ to k due to emission of phonon q . Similarly $W_{a,q}^{k+q \rightarrow k}$ refers to processes of absorption. The system is nonlinear, as the probabilities W depend on the product $f \cdot g$ of the electron and phonon distribution functions. The transfer of energy between electrons and phonons is due to the terms W , with a timescale of the order of 0.1 ps (see Fig. 5 below for more details). This equation set poses a multi-scale problem since the left hand sides involve different time scales: the velocity v_p of the phonons is two orders of magnitude lower than the velocity v_e of the electrons. Accordingly, the heat transfer by the lattice is much slower process than the charge transfer. Note that, when considering the electron lattice coupling, the energy transfer from electrons to the high-energy optical phonons is very efficient. However, optical phonons possess negligible group velocity and, thus, do not participate significantly in the heat diffusion.

They instead must transfer their energy to acoustic phonons, which diffuse heat. The energy transfer between phonons is relatively slow compared to electron-optical phonon transport and, thus, thermal non-equilibrium may also exist between optical and acoustic phonons. Fig. 5 shows the primary path of thermal energy transport and the associated time constants.

Fig. 5. The most likely path between energy carrying particles in a semiconductor device is shown together with the corresponding scattering time constants.

According to Fig. 5, the primary path of energy transport is represented first by scattering between electrons and optical phonons (T_{LO}) and then optical phonons to the lattice (T_A) (Tien *et al.*, 1998). The BTE for the two kinds of phonons is then used to provide the energy balance of the process. As we already mentioned, the direct solution of the phonon Boltzmann equation itself is a very difficult task as it is difficult to mathematically express the anharmonic phonon decay process, and in addition to this one has to solve separate phonon Boltzmann equation for each mode of the acoustic and optical branches. As already noted in Section 2 above, some attempts to solve the problem within the relaxation time approximation have been made by Narumanchi and co-workers (Naramunchi *et al.*, 2004). If we now include the electrons and holes into the picture with their corresponding Boltzmann transport equations, then the solution of the electron-hole-phonon coupled set of equations becomes a formidable task even for today's high performance computing systems. Therefore, some simplifications of this global problem are needed. Since for device simulation we are mainly focused on accurately calculating the *IV*-characteristics of a device, the self-heating being a by-product of the current flow through the device can be treated in a more approximate manner, but which is still more accurate than the local heat conduction model. Majumder and co-workers (Lai & Majumdar, 1996; Majumdar *et al.*, 1995), starting from the phonon Boltzmann equations derived energy balance equations separately for the optical phonon and the acoustic phonon bath. One can also derive these energy balance equations starting from the energy conservation principle. For electric field, $E \geq 10^6$ V/m, electrons lose energy to optical phonons and optical phonons decay to acoustic phonons. Using the first law of thermodynamics, the energy conservation equations for optical and acoustic phonons are:

$$\frac{\partial W_{LO}}{\partial t} = \left(\frac{\partial W_e}{\partial t} \right)_{coll} - \left(\frac{\partial W_{LO}}{\partial t} \right)_{coll}, \quad (17)$$

$$\frac{\partial W_A}{\partial t} = \nabla(\kappa_A \nabla T_A) + \left(\frac{\partial W_{LO}}{\partial t} \right)_{coll}, \quad (18)$$

W_e , W_{LO} and W_A are electron, optical phonon and acoustic phonon energy densities, respectively. We next use

$$dW_{LO} = C_{LO} dT_{LO} \text{ and } dW_A = C_A dT_A \quad (19)$$

where C_{LO} (specific heat capacity for optical phonons) can be estimated from Einstein model, while C_A (specific heat capacity for acoustic phonons) from Debye model. Next, the collision terms are expressed using the relaxation time approximation (RTA):

$$\left(\frac{\partial W_e}{\partial t} \right)_{coll} = n \cdot \frac{\frac{3}{2} k_B T_e + \frac{1}{2} m^* v_d^2 - \frac{3}{2} k_B T_{ph}}{\tau_{e-ph}}, \quad (20)$$

$$\left(\frac{\partial W_{LO}}{\partial t} \right)_{coll} = C_{LO} \frac{T_{LO} - T_A}{\tau_{LO-A}}, \quad (21)$$

where T_e is the electron temperature, v_d is the electron drift velocity and T_{ph} can be optical or acoustic phonon temperature depending on which kind of phonons electrons interact with. Combining Eqs. (17)-(21), one gets:

$$C_{LO} \frac{\partial T_{LO}}{\partial t} = \frac{3}{2} n k_B \left(\frac{T_e - T_{LO}}{\tau_{e-LO}} \right) + \frac{n m^* v_d^2}{2 \tau_{e-LO}} - C_{LO} \left(\frac{T_{LO} - T_A}{\tau_{LO-A}} \right) \quad (22a)$$

$$C_A \frac{\partial T_A}{\partial t} = \nabla(\kappa_A \nabla T_A) + C_{LO} \left(\frac{T_{LO} - T_A}{\tau_{LO-A}} \right). \quad (22b)$$

For electric field, $E < 10^6$ V/m, electrons lose energy directly to acoustic phonons and in that case:

$$C_A \frac{\partial T_A}{\partial t} = \nabla(\kappa_A \nabla T_A) + \left(\frac{\partial W_e}{\partial t} \right)_{coll} \quad (23a)$$

$$C_A \frac{\partial T_A}{\partial t} = \nabla(\kappa_A \nabla T_A) - \frac{3}{2} \frac{n k_B}{\tau_{e-A}} T_A + n \cdot \frac{\frac{3}{2} k_B T_e + \frac{1}{2} m^* v_d^2}{\tau_{e-A}} \quad (23b)$$

Under the assumption of very low electric fields, the electron temperature and acoustic phonon temperature equal the lattice temperature and terms 2 and 3 in Eq. (23) cancel. Using the low field conductivity and the mobility expressions we get that the heat source term reduces to the last term of Eq. (23b). Namely

$$q_{gen} = J \cdot E = \sigma E^2 = \frac{\sigma v_d^2}{\mu^2} = \frac{n m^* v_d^2}{\tau} \quad (24)$$

where it is assumed that for low doping concentrations the relaxation time τ that appears in Eq. (24) is the acoustic phonon relaxation time since acoustic phonon scattering, being isotropic scattering process, is most effective in randomizing the carrier momentum when the carrier energy is very low (low applied electric fields). In our electro-thermal simulator, we solve steady-state versions of the Eqs. (22a) and (22b) for the optical and acoustic phonon temperatures self consistently with the Monte Carlo simulation of the electron Boltzmann transport equation for modeling *n*-channel fully-depleted (FD) silicon on insulator (SOI) devices.

In other words, to simplify the calculations to a tractable form in the present work, we have used a Chapman-Enskog type expansion (Cercignani, 1988) to replace the microscopic phonon transport equation by a diffusion problem for the local density and energy of the phonons, where the diffusion coefficients are dependent on the state of the electrons. This method involves computation of the phonon energy dependent scattering tables in the Ensemble Monte Carlo (EMC) code for the electron transport, which already represents a big improvement over the current state of the art, where, as already discussed, the coupling of thermal and charge effects is strictly one way. This approach also takes care of the multi-scale nature of the problem, assuming a quasi steady state: the phonons are in steady state, albeit with a spatially dependent temperature distribution. In the present work we neglect the effect of the optical phonon dispersion on the tabulated scattering rates.

Using our electro-thermal simulator we demonstrate that, contrary to earlier predictions, the non-locality of the electron transport and velocity overshoot effects significantly reduce current degradation in nanoscale devices due to the fact that carriers traveling ballistically do not dissipate energy in the active region of the device, rather in the drain contact. We also show that in nanoscale devices, the hotspot corresponding to the peak temperature moves towards the drain end of the channel where removal of heat is more effective, and where less effect on the transport dynamics occurs underneath the gate.

4.1 Electro-Thermal Particle-Based Device Simulator Description

As illustrated in Fig. 6, and discussed in details in Ref. (Vasileska *et al.*, 2009), we self-consistently couple the Monte Carlo solution of the electron Boltzmann transport equation with the energy balance equations for both optical and acoustic phonons. This itself is a difficult problem as we are coupling a particle picture for electrons (which is inherently noisy) with a continuum model for the phonons. To achieve convergence of the coupled scheme, both temporal and spatial averaging of the variables extracted from the Monte Carlo (MC) solver (e.g. the electron density, drift velocity and temperature) must be performed. The number of simulated particles in the model contributes significantly to the smoothness of the variables being transferred to the energy balance solver. Within each "outer iteration" we solve the Boltzmann Transport equation for the electrons using the Ensemble Monte Carlo (EMC) method for a time period of 10 ps to ensure that steady state conditions have been achieved. The required variables are then passed to the thermal solver which gives the updated optical and acoustic phonon temperatures. This constitutes one Gummel cycle or "one outer iteration" (Raleva *et al.*, 2008; Sridhaharan *et al.*, 2008).

In our research effort, the EMC code for the carrier BTE solution (He, 1999; Ahmed, 2005) has been modified as well. As we have variable lattice temperature in the hot-spot regions, we have introduced the concept of *temperature dependent scattering tables*. For each combination of acoustic and optical phonon temperature, one energy dependent scattering

table is created. These scattering tables involve additional steps in the MC phase (Fig. 7 – right panel), because to choose randomly a scattering mechanism for a given electron energy, it is necessary to find the corresponding scattering table. To do that, first, the electron position on the grid needs to be found, in order to know the acoustic and optical phonon temperatures in that grid point, and then the scattering table with “coordinates” (T_L, T_{LO}) is selected. Using current state of the art computers, the pre-calculation of these scattering tables does not require much CPU time or memory resources and is done once in the initialization stages of the simulation for a range of temperatures. An interpolation scheme is then adopted afterwards for temperatures for which we do not have the appropriate scattering table.

Fig. 6. Flow-chart of the electro-thermal simulator.

Fig. 7. Left panel – Exchange of variables between the two kernels. Right panel: Choice of the proper scattering table.

To properly connect the particle-based picture of electron transport with continuous, “fluid-like” phonon energy balance equations, a space-time averaging and smoothing of electron density, drift velocity and electron energy are included. At the end of each MC time step, the electrons are assigned to the nearest grid point. Then, the drift velocities and thermal energies are averaged with the number of electrons at the corresponding grid points. After the MC phase, a time averaging of electron density, drift velocity and thermal energy is done and the electron temperature distribution is calculated. It is assumed that the drift energy is much smaller than the thermal energy. The smoothing of these variables is necessary, because most of the grid points, especially at the interfaces, are rarely populated with electrons. This leads to very low lattice temperatures in those points. The exchange of variables between electron and phonon solvers is shown on the left panel of Fig. 7.

Because SOI devices consist of two distinct regions, the silicon device layer and the buried oxide layer (in which the phonons have significantly smaller mean-free paths), the phonon BTE is solved in the silicon layer to accurately model heat transport, but the simpler heat diffusion equation is used in the amorphous BOX because the characteristic length-scale of conduction is much smaller than the film thickness. The two distinct computational regions are coupled through interface conditions that accounts for differences in material properties. For the coupling of the silicon and oxide solution domains, it is necessary to calculate the flux of energy through the interface between the two materials at each point along the interface for every time step.

To simulate the steady-state state behavior of a device, the system is started in some initial condition, with the desired potential applied to the contacts, and then the simulation proceeds in a time stepping manner until steady-state is reached. A common starting point for the initial guess is to start out with charge neutrality, i.e., to assign particles randomly according to the doping profile in the device, so that initially the system is charge neutral on the average. After assigning charges randomly in the device structure, charge is then assigned to each mesh point using an adequate particle mesh (PM) coupling method, and Poisson’s equation is solved. The forces are then interpolated on the grid, and particles are accelerated over the next time step. When the system is driven into a steady-state regime and MC simulation time has elapsed, we calculate the steady-state current through a specified terminal.

To continue with the thermal part of the simulation, the average electron density, drift velocity and electron temperature must be calculated on a grid. For the given bias conditions, the average electron density in the channel (Fig. 8) is very high at the Si-SiO₂ interface near the source injection barrier, while pinch off region exists near the drain.

Fig. 8. Sample electron density.

The two components of the electron kinetic energy are presented in Fig. 9. They show that the thermal part of electron kinetic energy is much larger than the drift part, so electron temperature can be calculated from the thermal energy.

Fig. 9. Left panel: electron thermal energy. Right panel: electron drift energy.

Finally the acoustic and optical phonon temperatures are calculated with the phonon energy balance equation solver. During the simulation, the gate contact and the bottom of the BOX are set to 300 K, while Neumann boundary conditions for the heat transfer are used in all other outer surfaces. Also, the tolerance used in the “thermal” successive over-relaxation (SOR) algorithm equals 0.001, which leads to very fast convergence.

When the simulation starts, all variables obtained from the first iteration of the EMC solver, are calculated using uniform distribution for the acoustic and optical phonon temperatures. This means that only one scattering table is used for all electrons, no matter where they are located in the device. When the phonon temperatures are computed from the phonon energy balance equations, they are “returned” at the beginning of the MC free-flight – scattering phase. Now, for each mesh point, we have a scattering table which corresponds to the acoustic and optical phonon temperatures at that point. In this case, the electron position defines which scattering table is “valid” and then, by generating a random number, the scattering mechanism is chosen for the given electron energy. The impact of the phonon temperature dependent scattering tables can be demonstrated by counting the number of

energy-exchange electron-phonon scattering events. From the results obtained it could be concluded that the inclusion of the phonon temperature dependent scattering table increases the number of electron-phonon interactions. We have also found that electrons with energies below 50 meV scatter mainly with acoustic phonons in silicon, while those with higher energy scatter strongly with the optical modes. The optical phonon modes have low group velocity (on the order of 1000 m/s) and their occupation is also relatively low, hence they contribute very little to the heat transport. The primary heat carriers in silicon are the faster acoustic phonon modes, which are significantly populated and have group velocities from 5000 m/s for transverse modes to 9000 m/s for longitudinal acoustic modes. Optical phonons decay into acoustic modes, but over relatively long time scales, i.e. picoseconds, compared to the order of tenths of picoseconds (Ferry, 2000). Under high field conditions, this can lead to the creation of a phonon energy bottleneck which can cause the density of optical phonon modes to build up over time, leading to more scattering events and impeding electron transport (Artaki & Price, 1989).

To test the overall convergence of the coupled EMC and thermal codes, we registered the variations of the drain current with the number of thermal iterations for a given bias condition. We have simulated 25nm fully-depleted SOI structure (Fig. 10 left panel). The device geometrical dimensions are: channel length = 25 nm, source/drain length=25 nm, Si layer thickness = 10 nm and BOX thickness = 50 nm. The results of simulations given in the right panel of Fig. 10, and show that only 3-5 iterations are necessary to obtain the steady-state solutions of the current.

Fig. 10. Left panel - Cross-section of the simulated devices. Right panel - Current decrease variations with the number of thermal iterations for different number of simulated electrons for 25nm fully-depleted SOI device. For the thermal part of simulations, the bottom of the BOX and the gate electrode are assumed to be isothermal boundaries, set to 300K. Substrate region is not modeled.

To further investigate the accuracy of the results and the convergence of the outer current iterative scheme, we increase the device width in order to increase the number of simulated electrons. Namely, the statistical error of the MC method shrinks as the number of simulated carriers, N_{sim} , increases, and it drops as $1/(N_{sim})^{1/2}$ (Ferry, 2000). The statistical

uncertainty of the results is 1.14% when device width is 1 μm and 0.65% for 3 μm device width, so the smoother convergence of the results is obtained with the larger number of simulated carriers (see Fig. 10 - right panel).

Simulation results for the isothermal and non-isothermal current are summarized in Table 1. It is obvious that SOI structure has larger current degradation, since SiO_2 has very poor thermal conductivity compared to diamond and AlN. But, even though diamond has over 6 times larger thermal conductivity than AlN, due to the differences in the dielectric constants, the current degradation is slightly higher in Si on AlN than in Si on Diamond. This is explained in more details in Ref. (Vasileska, et al., 2009) where we discuss the interplay of the dielectric constant of the BOX, the thickness of the BOX and the thermal conductivity of the BOX on the overall magnitude of the current and the current degradation for 25 nm channel length devices.

The left panel of Fig. 11 shows the effect of lattice heating on the I - V characteristics of a 25 nm gate length fully depleted SOI structure. The velocity overshoot is clearly seen on the right panel of Fig. 11. For this particular device structure, the corresponding degradation of device characteristics due to thermal effects is relatively small, less than 10%. As seen from the temperature maps of acoustic and optical phonons shown in Fig. 12, the maximum rise in the lattice temperature is on the order of 100 $^\circ\text{C}$ on the drain side of the gate as expected. For comparison, we also compare in Fig. 12 the effect of an elevated lattice temperature of 400K assuming an isothermal model compared to the non-uniform model, which shows that most of the effect observed in the I - V characteristic is due to lattice heating.

Device	Device width = 1 μm			Device width = 3 μm		
	Current (mA/ μm) isothermal	Average current (mA/ μm) thermal	Current decrease (%)	Current (mA/ μm) isothermal	Average current (mA/ μm) thermal	Current decrease (%)
FD-SOI	1.7706	1.6464	7.01	1.8238	1.6952	7.05
FD-SOAIN	1.7986	1.7669	1.76	1.8494	1.8201	1.58
FD-SOD	1.7884	1.7646	1.33	1.8398	1.8139	1.41

Table 1. Comparison of current degradation for different device technologies.

Fig. 11. Left panel: Output characteristics for $V_g = 1.2$ V. Top/middle curve correspond to the case of excluded/included lattice heating model. Bottom curve is an isothermal simulation but for lattice temperature $T = 400$ K throughout the whole simulation domain. Right panel:

Velocity along the channel for $V_G=1.2$ V and different values of V_D . Note that for $V_D>0.4$ V electrons are in the velocity overshoot regime which suggests that lattice heating does not significantly degrade the device characteristics.

Fig. 12. Left panel - Acoustic phonon temperature for $V_G=1.2$ and $V_D=1$ V. Notice the significant heating of the lattice that equals the acoustic phonon temperature in our model. Right panel - Optical phonon temperature for $V_G=1.2$ V and $V_D=1$ V. Notice the region near the drain with higher optical phonon temperature with respect to the acoustic phonon temperature.

4.2 Thermal degradation with scaling of device geometry

In addition to the previously noted observation regarding the influence of the velocity overshoot, we modeled larger fully-depleted SOI device structures and we also investigated the influence of the temperature boundary condition on the gate electrode on the current degradation due to heating effects. The geometrical dimensions of the simulated fully-depleted SOI MOSFETs are given in Table 2, while Table 3 gives the percentage of the current decrease due to heating effects with the variation of the gate electrode temperature for the device structures being considered. The calculated results show that the current degradation is more prominent for larger devices and for higher gate temperatures. For 80 nm and larger devices, simulated carriers are not in the velocity overshoot regime in the larger portion of the channel (especially near the source end of the channel). Snapshots of the lattice temperature profiles in the silicon layer for these devices when the gate temperature is set to 300K and 400K, are given in Fig. 13 and 14, respectively. From these snapshots one can observe that: (a) the temperature in the channel is increasing with the increase of the channel length, (b) the maximum lattice temperature region (hot spot) is in the drain and it shifts towards the channel for larger devices. This behavior is more drastic for higher gate temperatures (see Fig. 14).

Gate Length (nm)	Gate Oxide Thickness (nm)	Active Si-Layer Thickness (nm)	BOX Thickness (nm)	Channel Doping Concentration (cm ⁻³)	Bias conditions V _{gs} =V _{ds} (V)	Current (mA/μm) Isothermal value (300K)
25	2	10	50	1×10 ¹⁸	1.2	1.82
45	2	18	60	1×10 ¹⁸	1.2	1.41
60	2	24	80	1×10 ¹⁸	1.2	1.14
80	2	32	100	1×10 ¹⁷	1.5	1.78
90	2	36	120	1×10 ¹⁷	1.5	1.67
100	2	40	140	1×10 ¹⁷	1.5	1.57
120	3	48	160	1×10 ¹⁷	1.8	1.37
140	3	56	180	1×10 ¹⁷	1.8	1.23
180	3	72	200	1×10 ¹⁷	1.8	1.03

Table 2. Geometrical dimensions of the simulated fully-depleted SOI MOSFETs

Fig. 15 gives the ensemble averaged lattice and optical phonon temperatures along the channel in the silicon layer only for three technologies of devices being considered (25 nm, 80 nm and 180 nm). Notice that there is a bottleneck between the lattice and the optical phonon temperature in the channel which is more pronounced for shorter devices, due to the fact that the energy transfer between optical and acoustic phonons is relatively slow compared to the electron-optical phonon processes and the fact that the electrons are in the velocity overshoot (and since the channel is very short, they spent little time in the channel). To better understand the phonon temperature bottleneck, different cross-sections (at Si/SiO₂ interface, at half Si-layer width, at Si/BOX interface) of the lattice and the optical phonon temperature profiles in the channel direction were investigated as well. We find that the bottleneck is decreasing from Si/SiO₂ interface to Si/BOX interface. For shorter devices, it exists in the whole channel region, which is not a case for longer devices (thicker Si-layer and longer channel length). From the results we have presented here one can conclude that the higher the temperature in the channel and/or the longer the electrons are in the channel, the larger the degradation of the device electrical characteristics is due to the heating effects.

Type of simulation	Gate Electrode Temperature (K)	25 nm FD SOI nMOSFET $V_{gs}=V_{ds}=1.2V$		45 nm FD SOI nMOSFET $V_{gs}=V_{ds}=1.2V$		60nm FD SOI nMOSFET $V_{gs}=V_{ds}=1.2V$	
		Current (mA/um)	Current Decrease (%)	Current (mA/um)	Current Decrease (%)	Current (mA/um)	Current Decrease (%)
isothermal	300	1.929	\	1.5645	\	1.2457	\
thermal	300	1.8176	5.78	1.4717	5.93	1.1885	4.73
thermal	400	1.7467	9.45	1.3957	10.79	1.1124	10.83
thermal	600	1.5997	17.1	1.2612	19.39	0.9882	20.79

Type of simulation	Gate Electrode Temperature (K)	80 nm FD SOI nMOSFET $V_{gs}=V_{ds}=1.5V$		90 nm FD SOI nMOSFET $V_{gs}=V_{ds}=1.5V$		100 nm FD SOI nMOSFET $V_{gs}=V_{ds}=1.5V$	
		Current (mA/um)	Current Decrease (%)	Current (mA/um)	Current Decrease (%)	Current (mA/um)	Current Decrease (%)
isothermal	300	1.7810	\	1.6671	\	1.5743	\
thermal	300	1.5850	11	1.4805	11.19	1.4004	11.05
thermal	400	1.4612	17.96	1.3678	17.95	1.2833	18.48
thermal	600	1.2571	29.42	1.1784	29.31	1.1158	29.12

Type of simulation	Gate Electrode Temperature (K)	120 nm FD SOI nMOSFET $V_{gs}=V_{ds}=1.8V$		140 nm FD SOI nMOSFET $V_{gs}=V_{ds}=1.8V$		180 nm FD SOI nMOSFET $V_{gs}=V_{ds}=1.8V$	
		Current (mA/um)	Current Decrease (%)	Current (mA/um)	Current Decrease (%)	Current (mA/um)	Current Decrease (%)
isothermal	300	1.3657	\	1.2341	\	1.0321	\
thermal	300	1.1439	16.34	1.0416	15.6	0.8570	15.22
thermal	400	1.0630	22.16	0.9654	21.85	0.8100	21.52
thermal	600	0.9256	32.22	0.8311	32.27	0.6940	32.76

Table 3. Current decrease with gate temperature for different FD SOI technologies.

5. Open Problems

From the discussion presented so far, we might conclude that device simulation without lattice heating has reached high levels of sophistication even in the quantum domain area. The inclusion of lattice heating is done using the energy balance picture by the group from Arizona State University or by including the heat flux term that accounts for the number of generated acoustic and optical phonons in a given mode.

Fig. 13. Lattice temperature profiles in the silicon layer for FD SOI MOSFETs from Table 2 with gate temperature set to 300K. (25 nm -top, 100 nm-bottom).

Fig. 14. Lattice temperature profile in silicon layer for different device geometries of fully-depleted SOI MOSFET when gate temperature is set to 400K.

Fig. 15. Averaged lattice(full) and optical phonon(dashed) temperature profiles in the channel direction in the silicon layer for 25 nm (top), 80nm (middle) and 180 nm (bottom) FD SOI n-channel MOSFETs with gate temperature set to 300K.

Both of these approaches have their advantages and disadvantages. Important point is that understanding heat conduction surrounding nanostructures is just at the beginning. Many questions remain to be answered and new applications need to be explored. Some of these are discussed below.

1. The experimental studies carried out by Goodson's group provide the first quantitative evidence of non-local phonon heat conduction effects surrounding nanostructures. The effects are more pronounced at low temperatures, as expected (Sverdrup et al. 2001). In their experiments, the heater (also serving as a temperature sensor) is defined by a *pn*-junction. Because the calibration is through measuring the resistance of the heater when the sample temperature is uniformly raised, there are lingering questions on how accurate it is to use such calibration data to infer the heater temperature due to the spatial temperature gradient that may overlap with the space charge region of the *pn*-junction. More experimental studies are needed to provide further experimental data for analysis and to push to even smaller structures.

2. When the heat conduction is nonlocal, the transport is highly non-equilibrium, the temperature used to represent the modeling results is at best a measure of the local energy density, rather than their typical thermodynamic meaning. On the other hand, in microelectronics, the device reliability is often associated with the temperature through the Arrhenius law, which is a manifestation of the Boltzmann distribution and is a result obtained under the assumption of local equilibrium. A valid question is what does this temperature means for a device. If, for example, the calculated temperature corresponding

to the local energy density reaches the melting point, does it mean that local melting occur? The molecular dynamics simulations may shed some light on this question, as is discussed in Goodson et al.'s paper [Asheghi *et al.*, 1998].

3. The simulations so far are based on either Monte Carlo methods or the Boltzmann equation and take the various relaxation times as input parameters. These parameters are subject to a wide range of uncertainties. The phonon relaxation time, for example, were mostly derived from the modeling work in the 1950s and 1960s on the thermal conductivity of bulk materials. The relaxation time in these models is often obtained under various approximations, such as the Debye model for the phonon dispersion. There are many variables in these past models that can lead to different values of the relaxation time. There is a clear need for more accurate information on the relaxation times. Molecular dynamics simulation may be one way to obtain them. Similarly, electron-phonon scattering processes also need further consideration, particularly when electrons have very different temperatures from that of phonons.

4. In addition to the different application problems in microelectronics and in data storage, there are also new fundamental problems associated with nonlocal heat conduction. Past studies have considered phonons only. The concurrent electron and phonon nonlocal transport and nonequilibrium transport is an example. A recent proposal is to use the departure from equilibrium between electrons and phonons at a small contact point as a means to increase the thermoelectric energy conversion efficiency (Goshaei et al., 2002). The transport in this case can well become nonlocal for both electrons and phonons. This possibility was raised, but has never been studied (Chen, 1996).

5. The above examples emphasize the small length scales involved in nano-devices and nanomaterials. Short time scales are also becoming increasingly important. Similar questions can be raised for transport at short time scales as for the small length scales. Lasers can deliver a pulse as short as a few femtoseconds ($1 \text{ fs} = 10^{-15} \text{ s}$). Energy transduction mechanisms at such short time scales can differ significantly from that at macroscale. Microelectronic devices are pushing to the tens of gigahertz clock frequency with a much shorter transient time. The device temperature rise in such short time scales can be very different from predictions of the Fourier law.

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The main purpose of this book is to describe important issues in various types of devices ranging from conventional transistors (opening chapters of the book) to molecular electronic devices whose fabrication and operation is discussed in the last few chapters of the book. As such, this book can serve as a guide for identifications of important areas of research in micro, nano and molecular electronics. We deeply acknowledge valuable contributions that each of the authors made in writing these excellent chapters.

How to reference

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University Campus STeP Ri
Slavka Krautzeka 83/A
51000 Rijeka, Croatia
Phone: +385 (51) 770 447
Fax: +385 (51) 686 166
www.intechopen.com

InTech China

Unit 405, Office Block, Hotel Equatorial Shanghai
No.65, Yan An Road (West), Shanghai, 200040, China
中国上海市延安西路65号上海国际贵都大饭店办公楼405单元
Phone: +86-21-62489820
Fax: +86-21-62489821

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